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Description

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The present invention relates to refrigeration systems for air cooling a compartment and in particular, relates to air conditioning systems for vehicles. Systems of this type typically have refrigerant pumped by a compressor into an exothermic heat exchanger, or condenser, for liquification and expanded for flow through an endothermic heat exchanger, or evaporator, located in the compartment to be cooled and returned to the compressor in the form of gas or vapor. Vehicle passenger compartment air conditioning systems typically utilize a compressor energized by an electrical clutch connecting the compressor to the engine. In operation of modern vehicle air conditioning systems, it has been found necessary during low road speed operation of the ve-

hicle to provide cooling fan for directing a flow of ambient air over the condenser; and, it has further been de-10 sirable to provide such a fan which is electrically energizable independently of the vehicle engine speed.

In vehicular air conditioning systems employing an electrically clutched compressor and electrically operated condenser fan, it has been desired to provide suitable sensing or warning for declutching the compressor in the event of excessive sub-cooling of the liquid refrigerant, loss of refrigerant or an overpressure condition in the high pressure liquid side of the system.

Heretofore, protection against excessive sub-cooling and overpressure or loss of refrigerant has been provided by pressure sensors or transducers disposed in the liquid refrigerant line to detect the pressure in the line. These pressure sensing devices have proven to be costly and also have exhibited a history of reliability problems. It has become increasingly necessary to provide improved control of engine ignition, fuel feed and

load management in order to meet stringent emission requirements; and, therefore, the trend has been toward 20 all-electrical control of engine operating parameters, including engine operated accessories, such as air conditioning. In attempting to provide all electrical control of vehicle air conditioning, and particularly malfunction alarm and shut-down, it has been difficult to combine the pressure measurements with measurements of temperature in other portions of the refrigerant system for providing integrated electrical control of the entire sys-25 tem.

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Thus, it has been long desired to provide a way or means of providing overall control of the refrigerant system including high pressure and excessive sub-cooling warning by means of sensing only refrigerant temperature. This has proven to be difficult, inasmuch as sensing the actual temperature of the refrigerant does not permit the determination of the pressure in the refrigerant line. Thus, it has been desired to provide a way or means of electrically controlling a refrigerant system and providing warning in the event of excessive or low

pressure in the refrigerant without the necessity of providing pressure sensors.

Turning once more to the prior art, specific attention is drawn to US-A-3 293 876 which discloses a system of the type comprising the features of the first part of claim 1.

In accordance with the present invention a compressor cut-out and condensor fan cycling system as set 35 forth in the first part of claim 1 is characterized by the features of the second part of claim 1. The invention further provides for a method of controlling a refrigeration system as set forth in claim 10. It should be noted that with regard to claim 10 also said US-A-3 293 876 is of relevance as pointed out above. Preferred embodiments of the present invention are disclosed in the dependent claims.

SUMMARY OF THE INVENTION 40

The present invention provides a unique and novel way of sensing the condition of excessive or low refrigerant charge, and insufficient or excessive refrigerant pressure for providing an electrical control signal to cycle the condenser fan and disable the compressor clutch. The present invention employs a saturation temperature sensing thermistor disposed in the refrigerant line in the high side, or upstream of the expansion valve means. An electrical current is applied to the thermistor to heat the thermistor a sufficient amount to cause boiling of refrigerant on the surface thereof thus bringing the temperature of the thermistor to the saturation temperature of the refrigerant. A resistance is provided in series with the thermistor and the voltage drop across the resistor is measured, for a predetermined thermistor heating current, as the voltage in the thermistor

changes responsive to change in saturation temperature of the refrigerant. The thermistor thus measures saturation temperature which may be converted to saturation pressure from known properties of the refrigerant. During normal operation the fan is energized at a predtermined saturation temperature, and de-energized

when the temperature falls below a second predetermined saturation temperature.

When an excessive high pressure condition is detected, the microcomputer generates a signal to energize 55 the condenser fan and de-energize the compressor clutch. In another embodiment of the invention, a second thermistor disposed adjacent the heated thermistor for sensing the actual temperature of the liquid in a refriqerant line. Comparison of the actual temperature and the saturation temperature enables the amount of subcooling to be calculated; and, an electrical signal is generated for disabling the compressor clutch in the event of excessive or insufficient sub-cooling.

BRIEF DESCRIPTION OF THE DRAWING

Figure 1 is a schematic of an air conditioning system for a vehicle illustrating the placement of the heated thermistor for sensing saturation temperature of the liquid refrigerant on the high pressure side of the mechanical expansion valve means;

Figure 2 is a schematic of a portion of figure 1 illustrating an alternative capillary tube expansion means; Figure 3 is a schematic of a portion of the system for figure 1 illustrating an electrically controlled expansion valve;

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Figure 4 is a flow diagram of the electrical control signal generation for the system of figure 1; Figure 5 is an electrical schematic for the microprocessor based controller of figure 4;

Figure 6 is an electrical schematic of the circuit for conditioning the signal from the heated thermistor for the controller of figure 5; and,

Figure 7 is a portion of the schematic for an alternate embodiment of the invention employing a second thermistor in the high pressure line for sensing actual liquid refrigerant temperature.

DETAILED DESCRIPTION

- 20 Referring to Figure 1, the control system of the present invention is indicated generally at 10 as having a compressor 12 energized by an electrically operated clutch 14 powered by a belt 13 adapted to be driven from the vehicle engine. Refrigerant is pumped from the compressor 12 via conduit 16 through an exothermic heat exchanger or condensor 18 via conduit 20 to the inlet of an expansion means indicated generally at 22. In the embodiment of figure 1, the expansion means comprises a mechanically operated valve of the type employing
- a temperature responsive capsule operating a diaphragm for movement of the valve poppet, which valves are well known in the art. The low pressure discharge from the expansion valve means 22 flows through conduit 24 and through an endothermic heat exchanger or evaporator 26 and is evaporated and returned via conduit 28 through a passage in the valve 22 and conduit 30 to the inlet of the compressor.
- A thermistor T_s is disposed in the conduit 20 adjacent the inlet of the expansion means 22 for sensing the saturation temperature of the liquid refrigerant entering the expansion valve. An electrically operated blower or fan 32 is disposed adjacent the condenser 18 for directing a flow of ambient air thereacross upon electrical energization of the fan motor 34. One power lead 36 of the motor 34 is grounded and the remaining lead 38 is connected to the power output of a fan relay 40. The remaining power lead 42 of the fan relay is connected to the vehicle 12 volt supply via junction 44.
- ³⁵ Junction 44 also is connected to the input of a voltage regulator 48 (via lead 46) which supplies power to the microprocessor based controller 50. Controller 50 receives signal inputs from comparator 52 and provides a signal output to the fan relay 40 along line 54 and switches power to the compressor clutch 14 along line 56. Saturation thermistor T_s has applied thereto via lead 58, a voltage from the voltage regulator indicated generally +V, junction 60 and lead 62. The remaining lead from 64 of T_s is grounded through calibration resistor
- 40 R33. The voltage drop across resistor R33 is detected at junction 66 and is applied through lead 68 to the comparator 52.

A suitable line power enabling switch 70 is provided for applying vehicle battery power to junction 44. The switch 70 is adapted for remote selective actuation by the vehicle operator, as for example by instrument or dash panel mount, for initiating operation of the refrigeration system.

In the presently preferred practice, T_S is an NTC thermistor manufactured by Fenwall Electronics, 63 Fountain Street, Farmingham, Massachusetts 01701 and has identification FD21J1-W and has a resistance of 100 ohms at 25°C.

Referring now to Figure 2, an alternate form of the system of figure 1 is illustrated wherein the refrigerant expansion means 222 comprises a coiled capillary tube. Referring to Figure 3, another embodiment of the invention of figure 1 is illustrated wherein the expansion means comprises an electrically operated expansion valve 322 which may be of the type having a solenoid energized by a pulse-width-modulated control signal such as that described in co-pending application serial number 007,147 filed January 27, 1987 assigned commonly to the assignee of the present application. The conduit connections and the disposition of the saturation temperature sensor T_s are otherwise the same for the embodiments of figures 2 and 3 as the embodiment of

55 figure 1.

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Referring now to figures 4, 5 and 6, closure of the switch 70 enables operator selective initiation of the system to energize the compressor at step 72. When a suitable time delay has occurred, typically three seconds, the controller 50, is operable to energize the saturation thermistor T_s at step 76. The controller 50 then

monitors the voltage drop across resistor R33 and compares the voltage V_R across resistor R33 with a reference voltage typically 6.6 volts in the comparison step 78. When a suitable time delay, preferably three-fourth second, has occurred at step 80, the comparison function is repeated. Otherwise, if V_R is greater than 6.6 volts, the condenser fan motor 34 is energized at step 82. When the voltage drop V_R rises to a predetermined limit,

- 5 typically 7.3 volts, the comparator is operative at step 84 to de-energize the compressor clutch and the fan motor at step 86. If an electrically operated valve such as valve 322 (figure 3) is employed in the system of figure 1, step 86 is also operable to open or maintain open the valve 322 to permit system pressure equalization after shut down. If the comparison at step 84 determines that V_R has not exceeded the limit of 7.3 volts and so long as V_R does not drop below a predetermined lower limit, typically 5.7 volts, the compressor will remain energized; and, after the three quarter second time delay in step 80, the comparison of step 78 will be repeated.
 - energized; and, after the three quarter second time delay in step 80, the comparison of step 78 will be repeated. In the event that the voltage drop V_R falls below the lower limit of 5.7 volts, as determined by comparison step 88, the controller 50 is then operable in step 90 to disable power to the condenser fan motor 34.

In the event that V_R has been detected in step 84 as greater than the upper limit 7.3 volts, and the system has been de-energized in step 86, and after a suitable time delay determined in step 92, typically 20 seconds, the controller is operable to re-energize the compressor clutch at step 72.

Referring to figure 5, the microprocessor U4, which in the presently preferred practice comprises a solid state device bearing manufacturer's designation 6805P2 available from Motorola Semiconductor Products, Algonquin Road, Schaumberg, Illinois 60195, receives power at pin 3 thereof from the voltage regulator 48. The regulator preferably comprises a solid state device U1 bearing manufacturer's designation MC7805 available from the said Motorola, which receives at pin 1 thereof, a voltage V_B from the vehicle battery, typically nine

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- to sixteen volts, through diode CR1 and resistor R1. Protective devices comprising capacitor C1 and zener diode CR6 in parallel protect the input from transient spikes and are grounded along with pin 2 of U1. A capacitor C2 is disposed across the output pin 3 of U1, which output gives a regulated five volts to the input at pin 3 of the processor U4. The processor U4 is operative on power up to provide at output pin 12 a signal through resistor R28 to Q12, which is biased by the vehicle battery voltage V_B through resistor R25 at its collector. When
 - the signal from pin 12 of the processor U4 switches Q12 OFF and the voltage V_B is applied to base of Q7. Referring to figure 6, when the voltage V_B is received at the base of Q7, Q7 conducts and drops the voltage at junction 94 which is biased through resistor R23 positive from the voltage regulator 48. When the voltage at junction 94 is dropped, the base of Q6 is biased low through resistor R26, and Q6 is caused to conduct current
- from junction 96 which is biased positive by voltage from regulator 48 and also a voltage from junction 98. Current flowing through Q6 flows through thermistor T_S and calibration resistor R33 to ground. T_S is thus heated to a temperature sufficient to boil the liquid refrigerant flowing thereover by a limited current therethrough; and, variations in the temperature are sensed by detecting changes in the voltage drop across resistor R33 as V_R measured at junction 66. V_R is applied to junction 102 and through resistor R17 to the positive input pin 11 of
- ³⁵ device U2 which comprises in the presently preferred practice a dual comparator device bearing manufacturer's designation LM339 available from National Semiconductor Corp., 2900 Semiconductor Drive, Santa Clara, California 95051. The negative input pin 10 of U2 receives a voltage through resistor R24 from junction 104 which is biased by a positive voltage V_z from regulator 48 and also battery voltage through resistor R16 from junction 106 which is connected to ground through reverse biased diode CR11. A resistor R24 is connected to
- ⁴⁰ pin 10 of U2 and is grounded such that R24 and R32 comprise a divider to apply a regulated reference voltage to pin 10 of U2. When V_R exceeds 6.6 volts, the output of comparator U2 at pin 13 goes high and is applied to junction 108 which is biased by a voltage from the regulator 48 through resistor R18. The voltage at junction 108 is applied through resistor R22 to the base of Q8 such that when the output of U2 is high Q8 is turned ON and the output thereof applied to collector junction 110, which is biased through resistor R19 at five volts and is applied to pin 9 of the microprocessor U4.
 - Referring to figure 5, when the input to pin 9 of U4 goes low, the output at pin 17 provides a signal through resistor R13 to the base of Q4 which turns OFF leaving a positive bias of the battery voltage through resistor R12 to the collector terminal thereof, which voltage is applied to the base of power FET Q11, causing Q11 to conduit at its output pin 2 to junction 112 which is connected to fan relay 40 along lead 54. The output of Q11
- 50 at junction 112 is protected by zener diode CR9 and diode CR12 which are grounded along with pin 3 of Q11. The output of Q4 is similarly protected at pin 1 of Q11 by diode CR14A which is switched to ground by Q5 which has its base biased through resistor R15 and R14 by a voltage from the regulator 48. A junction 114 between resistors R14 and R15 is signaled by an output at pin 19 of U4 to turn Q5 ON to ground the output of Q4 during power up.
- ⁵⁵ Referring to figures 5 and 6, when V_R at pin 11 of U2 is less than a lower limit voltage level, typically 5.7 volts, Q8 is turned OFF and the voltage at pin 9 of U4 goes high (positive 5 volts) and the output at pin 17 turns Q4 ON grounding the output thereof and turning Q11 OFF to cut off the voltage to the fan relay thereby disabling fan motor 34.

Referring to figure 6, V_{R} at junction 102 is also applied to negative pin 4 of another half of device U2; and, the positive input of comparator U2 at pin 5 thereof, is connected to junction 116 which is grounded through resistor R21 and receives a positive bias Vz through R13. The output of comparator U2 at pin 2 is applied to junction 118 which receives a positive voltage bias through resistor R14 from the voltage regulator 48.

Referring to figure 5, the output of U2 at junction 118 denoted HPCO in figure 5 is applied through resistor R30 to the base of Q14 which has its emitter grounded and its collector output connected through junction 118. Junction 118 is biased through resistor R27 by a positive five volts from regulator 48 and is connected to input pin 15 of microprocessor U4.

Referring to figure 6, the input to pin 5 of comparator U2 from junction 116 is maintained at a positive regulated voltage, typically 7.3 volts by the voltage divider resistors R19, R21. When the voltage on pin 4 drops below the voltage on pin 5, namely 7.3 volts, the output of U2 goes high turning Q14 OFF which causes a positive 5 volts from junction 118 to be applied to pin 15 of U4 and the output at pin 18 of U4.

Referring to figure 5, device CY1 is an oscillator in series to ground with capacitor C13 and in parallel with capacitor C12. CY1 is connected to pins 5 and 4 of microprocessor U4 for providing timing pulses preferably on the order of 4 megahertz.

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28 of the microprocessor U4.

Controller 50 includes a "dead man" timer which monitors the microprocessor function. U4 provides an output on pin 14 thereof through capacitor C7 to junction 119 which is connected to the base of Q1 and to ground thtough resistor R7. Q1 has the emitter grounded and the collector junction connected to junction 120 which is connected to pin 6 of device U5 which comprises a Fenwall NE555D timer. U5 also has pin 6 connected to

junction 120, pin 5 grounded through capacitor C9 and pin 1 grounded. Pin 7 of U5 is connected to junction 20 122 which is biased through resistor R3 with a positive voltage from the regulator 48; and, the voltage of junction 122 is also applied through resistor R6 to the collector junction of Q1. Pin 4 of U5 is connected to junction 124 which is grounded through capacitor C10 and junction 124 also is biased by a positive voltage through resistor R4. The output at pin 3 of U5 is connected through capacitor C11 to junction 126, which is biased through resistor R5 by positive five volts from the regulator 48, and is protected by diode CR5 and is connected to input

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The device U5 is operative such that if a signal is not received from pin 14 to Q1, after 70 milliseconds, U5 is not reset by Q1 and provides a reset signal through pin 28 to the microprocessor U4.

When the microprocessor is operative to provide an output signal at pin 18, the signal is applied through resistor R11 to the base of Q3 which has its emitter grounded. The collector junction of Q3 is biased to a positive 30 voltage from the battery through resistor R10 and is connected through junction 128 to the base or pin 1 of power FET device Q10. The output at pin 2 of Q10 is connected to the compressor clutch; and, the output at pin 3 of Q10 is grounded, with the outputs protected by zener diode CR8 connected to pin 2 and diode CR10 connected to output pin 3. When a signal is received from pin 18 of U4 to the base of Q3, Q3 conducts thereby 35 dropping the voltage on junction 128 and turning power FET Q10 OFF thereby de-energizing the compressor

clutch.

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Similarly, an output from pin 16 of U4 is applied through resistor R9 to the base of device Q2 which has its emitter grounded and its collector junction biased to a positive voltage from the battery through resistor R8 and connected to junction 130. Junction 130 is connected to input pin 1 or base of power FET Q9 which has its output pin 3 grounded and output pin 2 connected to the coil of refrigerant valve 322 for the embodiment of figure 3. The outputs are protected by zener diodes CR7, CR4 and diode CR16. When a signal is received from U4 pin 16 to the base of Q2, Q2 conducts, thereby dropping the voltage to junction 130 and turning Q9

OFF, thereby disabling the valve coil for turning the valve OFF.

- Referring now to figure 7, the control system of the present invention is shown with an optional thermistor 45 T_A disposed in the liquid refrigerant line closely adjacent the saturation temperature thermistor T_S for sensing the actual temperature of the liquid refrigerant in the line. With reference to figure 5, T_A receives a positive battery voltage through a diode with the remaining lead connected to a junction 130 which is connected through capacitor C5 to ground and also to input pin 2 of a device U3. In the presently preferred practice U3 comprises a Fenwall NE5560 timer which receives a positive five volts from the regulator 48 at pin 4, with pin 7 grounded
- 50 and pin 3 grounded through capacitor C15. A signal from the microprocessor pin 10 is applied to pin 6 of U3 and triggers U3 to discharge capacitor C5. When the voltage on pins 1 and 2 of U3 from junction 130 reach two-thirds of the bias voltage, U3 applies a signal through its output pin 5 and diode CR12 to junction 132 which is connected to input pin 2 of U4 and also grounded through resistance R2.

The microprocessor U4 measures the time to receive the signal, the time measurement giving a digital 55 representation of the voltage on T_A . The microprocessor can then look up the temperature of T_A from a table of voltages and resistances provided by the manufacturer of the thermistor T_A. In the presently preferred practice, T_A is a 30 kohm NTC thermistor available from Fenwall bearing manufacturer's designation UUR43J21.

The actual temperature sensed by T_A can then be employed to determine the sub-cooling of the system

according to the equation

"Sub - Cooling = $T_S - T_A$ ".

When the sub-cooling preferably exceeds more than 75 degrees, this indicates that there is an overcharge.
If the sub-cooling is zero, this provides an indication that there is insufficient refrigerant in the system to provide
proper cooling; and, the operator can be alerted to turn off the system in either case. The microprocessor then provides a signal on output pins 17 and 18 to de-energize the condenser fan and the compressor clutch to disable the system. In the event that an electrically operated expansion valve such as valve 322 of figure 3 is employed, the microprocessor would also provide an output signal at pin 16 to disable the valve coil. This is accomplished by a signal through R9 to the base of Q2 which conducts to ground the voltage bias from the
battery through R8 at its collector which is connected to junction 130 and base of power FET Q9. Q9 is then caused to turn OFF cutting current flow to the valve coil. The output of Q9 is protected by zener CR7 and diodes

CR4 and CR16.

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The values of the resistors, capacitors and diode designations are given in table I.

20	Resistance OHM	Capacitances <u>Microfarads</u>	Diodes and Other
	R1-56 R2-10K R3-100K	Cl 10, 35V C2 .1, 50V	CR1 GL41D CR2 MMBD914 CR3
25	R4-30K R5-10K R6-1K	CR-1 C51	CR4 IN5349 12V, 5W CR5 MMBD 914 CR6 MLL 4145, 19V, 1W
	R7-1K R8-1K1 R9-2.2K R10-1K	C701 C81 C901	CR7 IN5349, 12V, 5W CR8 IN5352, 24V, 5W CR9 MLL4745
30	R11-2.2K R12-1K R13-2.2K	C10-2.2 C11-2.2 C12-18 pico C13-18 pico	U1 MCT805 U2 LM 339 U3 NE556D U4 M6805P2
35	R14-10K R15-2.2K R16-100 R17-10K	C1501	U5 NE555D Q1 2N3904 Q2 2N3904
40	R18-2.7K R19-4.7K R20-6.8K R21-7.3K	C171 C181	Q3 2N3904 Q4 2N3904 Q5 2H3304 Q6 2N6710 Q7 2N3004
	R22-10K R23-10K R24-12K		Q7 2N3904 Q8 2N3904 Q9 BFS130 Q10 BFS130
45	R25-4.7K R26 470 R27-4.7K R28-2.2K		Q11 BFS 130 Q12 2N3904 Q14 2N3904 CR10 1N4002
50	R29-2.7K R30-2.2K R31-2.2K R32-2.2K		CR11 10V, 1W CR12 SL411 CR13 MMBD914 CR14 MMBD283B
	R33-13		CR15 MLL4346, 18V, 1W CR16 GL41D

TABLE I

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The procedure for determining the value of the calibration resistor R33 will now be described wherein the thermistor T_s is placed in the desired location in the refrigerant line and a digital pressure transducer is disposed temporarily in the system and the system is operated with logging of the pressure transducer output and the

voltage drop V_R . The data logger is fed into a computer which is programmed to perform the calculation to determine the resistance of the thermistor from the following equation:

$$R_t = R(V - V_r/V_r)$$

where R_t is the resistance of T_s , R is the resistance of R33 and V is the supply voltage.

The computer then calculates the temperature of the thermistor T_s by using the temperature/resistance values supplied by the thermistor manufacturer and which are set forth in Table II below:

	TABLE I	.1
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15	o _F	°C	R-T <u>Curve</u>	Alpha Temp. <u>Coeff.</u>	Resis. _Dev
10	-76	-60	49.10	6.0	0 7
	-58	-50	27.54		9.7
	-40	-40	16.08	5.6	8.2
	-22	-36		5.2	6.8
	-4		97.03	4.9	5.5
20		-20	6.053	4.5	4.4
	14	-10	3.890	4.3	3.3
	32	0	2.568	4.0	2.3
	50	10	1.731	3.8	1.3
	68	20	1.194	3.6	0.3
25	77	25	1.00	3.5	0.0
20	86	30	.8413	3.4	0.6
	104	40	.8040	3.2	1.4
	122	50	.4412	3.1	2.4
	140	60	.3275	2.9	3.1
	158	70	.2468	2.8	3.7
30	176	80	.1856	2.7	
	194	90	.1460	2.6	4.4
	212	100			5.1
	4a ± 4a	100	.1140	2.5	5.7

R-J: multiply resistance at 25°C by listed valve to obtain resistance at temperature.
 Alpha temperature coeff: denotes precent in resistance change per °C at a specific temperature.
 Resistance Deviation: add to resistance tolerance at reference temperature (250°C) to give complete percentage of resistance deviation.

The computer then calculates the equivalent system pressure based on the assumption that the heated thermistor has measured saturation temperature. The saturation pressure/temperature relationship is known and is defined for each refrigerant by the various manufacturers; and, for freon 12 the following relationship is used:

 $P = 9.1473 + .50688 * T + 4.13189609 * T^2 + 1.451827638E - 5 * T^3 + 1.14699897E - 8 * T^4;$ where P = the pressure (psig) and T = the saturation temperature (degrees F).

45 The thermistor temperature based upon resistant measurement may also be computed from the following

formula:

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 $T = (3508.96) - \log (100/R_T) * 1.1 + 11.79611) - 273.15) * 1.8 + 32;$

Where T = temperature of the thermistor (degrees F) and R_T = resistance of the thermistor ohms.

- The actual pressure obtained from the calibration pressure transducer output and the calculated system pressure derived from T_s are plotted against time as the system loads are randomly varied. If the pressure correlates throughout the test, then the value of the limit resistor R33 selected is correct. If the calculated system pressure is lower than the actual transducer, then the value of the resistor R33 is too high which unduly limits the wattage to the thermistor T_s and fails to maintain surface boiling of the refrigerant contacting the thermistor. If the calculated pressure is higher than the actual pressure measured by the transducer, then the
- value of the resistor R33 is too low, which means excessive wattage (current) has been applied to the thermistor. The process of iteration is continued with new values of R33 substituted until the pressure calculated correlates with the actual pressure measured under the range of service conditions anticipated. It has been found that the final selection of resistor R33 yields a 60 psig offset in pressure between the actual pressure

measured by the transducer and the pressure calculated from T_s over the range of expected service conditions. This offset can be nulled or zeroed out of the software to provide the desired correlation over the operating range of the system.

When the value of the resistor R33 has been determined a correlation can be made between the voltage V_R across the current limiting resistor R33 and the saturation temperature derived from T_S . This correlation of values can be programmed into microprocessor U4 such that T_S can be found for a given input for V_R .

The present invention thus employs a novel technique or way of determining whether a condition of overpressure, under-pressure, overcharge or insufficient charge of refrigerant is present in an automotive air conditioning system by the use of inexpensive thermistors in the refrigerant line and without the need of costly

- 10 pressure transducers. The system of the present invention employs a heated thermistor in the liquid refrigerant line on the high pressure side of the expansion means, or expansion valve, which enables a calculation to be made of the saturation pressure based upon the assumption that saturation temperature has been measured. An optional auxiliary thermistor is employed to measure the actual temperature of the refrigerant and a comparison of the saturation temperature measured by the heated thermistor with the actual temperature is used
- 15 to determined the degree of sub-cooling from which a condition of normal or abnormal refrigerant charge may be deduced. In the event of abnormal refrigerant charge, the system is alarmed to automatically disable the refrigerant compressor and condenser fan. During normal operation, the heated thermistor is employed for sensing saturation temperature and a derived saturation pressure enables the system to determine whether a normal or abnormal condition of refrigerant pressure exists; and, in the event of an abnormal, either too high
- 20 or too low refrigerant pressure condition, the system is automatically operative to disable the compressor and enable condenser fan and to reenable the compressor when the abnormal pressure condition is dissipated. The present invention thus provides a simple and economical refrigeration control system employing only

thermistors in the refrigerant line to provide alarm notification of high or low refrigerant pressure and optionally high or low refrigerant charge and automatically disable the compressor and condenser fan. The control system

of the present invention may be employed with a refrigerant expansion means comprising any of three types; namely, a simple capillary tube, a mechanical pressure/temperature expansion control valve employing a pressure sensitive diaphragm for controlling valve poppet movement or an electrically operated expansion valve.

30 Claims

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1. A compressor cut-out and condensor fan cycling system for controlling refrigerant between an exothermic and an endothermic heat exchanger comprising:

pump means operable upon energization to compress said refrigerant and discharge same under pressure;

conduit means (20) operatively connecting the discharge of said pump means through said exothermic heat exchanger and said endothermic heat exchanger and return to the intake of said pump means to said endothermic heat exchanger;

expansion means (22) disposed in said conduit means and operative to control flow of refrigerant between said exothermic heat exchanger and said endothermic heat exchanger;

a temperature sensing means disposed in said conduit between said exothermic heat exchanger and said expansion means for sensing the temperature of said refrigerant on the high pressure side of said expansion means;

fan means operable upon electrical energization to direct a flow of cooling air over said exothermic heat exchanger; and

controller means (50) operable in response to said temperature detection to energize said fan means when said detected temperature is greater than a first predetermined level and operable to deenergize said pump means when said detected temperature is greater than a second predetermined level; characterized in that

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said temperature sensing means comprises a saturation temperature sensing thermistor; resistance means is in series electrically with said thermistor;

circuit means is operable to effect a flow of current through said resistance means and said thermistor and including means operable to detect the voltage drop across said resistance means upon said flow of current;

and said controller means (50) is operable in response to said voltage detection.

2. The system defined in claim 1, wherein said expansion means (22) comprises an electrically operated expansion valve.

- 3. The system defined in claim 1, wherein said expansion means comprises a capillary tube.
- 4. The system defined in claim 1, wherein said expansion means comprises a temperature and pressure responsive diaphragm control valve.

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- 5. The system defined in claim 1, wherein said first predetermined voltage level is established at about six and six-tenths (6.6) volts.
- **6.** The system defined in claim 1, wherein said second predetermined level is established at about seven and three-tenths (7.3) volts.
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- 7. The system defined in claim 1, wherein said controller means (50) is further operative to de-energize said fan means when said voltage drop across said resistance means is less than said first level by a predetermined differential.
- **8.** The system defined in claim 1, wherein said controller means (50) is further operative to re-energize said pump means at a predetermined time interval after de-energization.
 - **9.** The system defined in claim 1, further comprising:
 - (a) a second thermistor disposed in said conduit at generally the same station as said saturation temperature sensing thermistor;
 - (b) circuit means for providing a voltage signal indicative of the change in resistance of said second thermistor responsive to changes in the actual saturation temperature of said refrigerant; and,
 - (c) means operable to compare said second thermistor voltage signal with said detected voltage and de-energize said pump means when said comparison exceeds a third predetermined value or is less than a fourth predetermined value.
 - **10.** A method of controlling a refrigeration system of the type having a pump operable upon energization for circulating refrigerant between an exothermic heat exchanger and endothermic heat exchanger and a fan for directing air flow over said exothermic heat exchanger comprising the steps of:
- (a) providing expansion means (22, 222) for controlling flow of refrigerant from said exothermic heat exchanger to said endothermic heat exchanger; and,
 (b) providing a sensor (T_s) for sensing the temperature of refrigerant entering said expansion means
 - and connecting a resistance electrically in series with said sensor;
 - (c) flowing an electrical current through said resistance and said sensor;
- (d) detecting the voltage drop across said resistance when said current is flowing; and
 (e) energizing a fan (32) and directing a flow of cooling air over said exothermic heat exchanger when said detected voltage drop is greater than a first predetermined value; and,
 (f) de-energizing said pump means when said detected voltage drop is greater than second predetermined value.
 - **11.** The method defined in claim 10, wherein said step of providing a sensor includes the step of providing a thermistor (T_s) in the flow of refrigerant to said expansion means.
 - **12.** The method defined in claim 10, wherein said step of energizing said fan includes the step of comparing said detected voltage with a reference voltage.
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- **13.** The method defined in claim 10, wherein said step of de-energizing said pump means includes the step of comparing said detected voltage with a reference value.
- **14.** The method defined in claim 10, further comprising the step of de-energizing said fan when said detected voltage drop is greater than a second predetermined value less than said first predetermined value.

Patentansprüche

 Kompressortrenn- und Kondensatorgebläsezyklus-System zum Steuern eines Kühlmittels zwischen einem exothermen und einem endothermen Wärmetauscher, das folgendes aufweist: Pumpenmittel, die betriebsmäßig auf eine Erregung hin das Kühlmittel komprimieren und dasselbe unter

Druck abgeben;

Leitungsmittel (20), die betriebsmäßig den Auslaß der Pumpenmittel durch den exothermen Wärmetauscher und den endothermen Wärmetauscher und zurück zu dem Einlaß der Pumpenmittel mit dem endothermen Wärmetauscher verbinden;

- Expansionsmittel (22), die in den Leitungsmitteln angeordnet sind und betriebsmäßig die Strömung von Kühlmittel zwischen dem exothermen Wärmetauscher und dem endothermen Wärmetauscher steuern; Temperaturabfühlmittel, die in der Leitung zwischen dem exothermen Wärmetauscher und den Expansionsmitteln angeordnet sind zum Abfühlen der Temperatur des Kühlmittels auf der Hochdruckseite der Expansionsmittel;
- Gebläsemittel, die betriebsmäßig auf eine elektrische Erregung hin eine Strömung von Kühlluft über den exothermen Wärmetauscher lenken oder richten; und Steuermittel (50), die betriebsmäßig ansprechend auf die Temperaturdetektion die Gebläsemittel erregen, wenn die detektierte Temperatur größer ist als ein erstes vorbestimmtes Niveau und betriebsmäßig die Pumpenmittel enterregen, wenn die detektierte Temperatur größer ist als ein zweites vorbestimmtes
 Niveau;
 - dadurch gekennzeichnet, daß:

die Temperaturabfühlmittel einen die Sättigungstemperatur abfühlenden Thermistor aufweisen; Widerstandsmittel, elektrisch in Serie mit dem Thermistor geschaltet sind;

- Schaltungsmittel, betriebsmäßig einen Stromfluß durch die Widerstandsmittel und den Thermistor bewir ken und Mittel aufweisen, die betriebsmäßig den Spannungsabfall an den Widerstandsmitteln auf einen
 Stromfluß hin detektieren; und die Steuermittel (50) betriebsmäßig auf die Spannungserkennung oder
 Detektion ansprechen.
 - 2. System nach Anspruch 1, wobei die Expansionsmittel (22) ein elektrisch betätigtes Expansionsventil aufweisen.
 - 3. System nach Anspruch 1, wobei die Expansionsmittel ein Kapillarrohr aufweisen.
 - 4. System nach Anspruch 1, wobei die Expansionsmittel ein auf Temperatur oder Druck ansprechendes membranbetätigtes Steuerventil aufweisen.
 - 5. System nach Anspruch 1, wobei das erste vorbestimmte Spannungsniveau bei ungefähr sechs und sechszehntel (6,6) Volt eingestellt ist.
- 6. System nach Anspruch 1, wobei das zweite vorbestimmte Niveau auf ungefähr sieben und dreizehntel
 ³⁵ (7,3) Volt eingestellt ist.
 - 7. System nach Anspruch 1, wobei die Steuermittel (50) ferner betriebsmäßig die Gebläsemittel enterregen, wenn der Spannungsabfall an den Widerstandsmitteln geringer ist als das erste Niveau und zwar um eine vorbestimmte Differenz.
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- 8. System nach Anspruch 1, wobei die Steuermittel (50) ferner betriebsmäßig die Pumpenmittel wieder erregen und zwar nach einem vorbestimmten Zeitintervall nach der Enterregung.
- 9. System nach Anspruch 1, wobei das System ferner folgendes aufweist:
- (a) einen zweiten Thermistor, der in der Leitung angeordnet ist, und zwar im allgemeinen an derselben Stelle wie der die Sättigungstemperatur abfühlende Thermistor;
 (b) Schaltungsmittel zum Vorsehen eines Spannungssignals das die Widerstandveränderung des zweiten Thermistors ansprechend auf die Veränderungen der tatsächlichen Sättigungstemperatur des Kühlmittels anzeigt; und
- 50 (c) Mittel, die betriebsmäßig das zweite Thermistorspannungssignal mit der detektierten Spannung vergleichen und die Pumpenmittel enterregen, wenn der Vergleich einen dritten vorbestimmten Wert übersteigt oder geringer ist als ein vierter vorbestimmter Wert.
 - 10. Verfahren zum Steuern eines Kühlsystems der Bauart mit einer Pumpe, die betriebsmäßig auf eine Erregung hin Kühlmittel zwischen einem exothermen Wärmetauscher und einem endothermen Wärmetauscher zirkuliert, und mit einem Gebläse zum Richten oder Lenken einer Luftströmung über den exothermen Wärmetauscher, wobei das Verfahren die folgenden Schritte aufweist:
 - (a) Vorsehen von Expansionsmitteln (22, 222) zum Steuern der Kühlmittelströmung von dem exother-

men Wärmetauscher zu dem endothermen Wärmetauscher; und

(b) Vorsehen eines Sensors (T_s) zum Abfühlen der Temperatur des Kühlmittels, das in die Expansionsmittel eintritt, und elektrisches in serieschalten eines Widerstandes mit dem Sensor;

(c) Bewirken eines elektrischen Stromes durch den Widerstand und den Sensor;

(d) Detektieren des Spannungsabfalls an dem Widerstand wenn der Strom fließt; und (e) Erregen eines Gebläses (32) und Lenken oder Richten einer Kühlluftströmung über den exothermen Wärmetauscher, wenn der detektierte Spannungsabfall größer ist als ein erster vorbestimmter Wert; und

(f) Enterregen der Pumpenmittel, wenn der detektierte Spannungsabfall größer ist als der zweite vorbestimmte Wert.

- **11.** Verfahren nach Anspruch 10, wobei der Schritt des Vorsehens eines Sensors den Schritt des Vorsehens eines Thermistors (T_S) in der Kühlmittelströmung zu den Expansionsmitteln aufweist.
- **12.** Verfahren nach Anspruch 10, wobei der Schritt des Erregens des Gebläses den Schritt des Vergleiches der detektierten Spannung mit einer Bezugsspannung aufweist.
 - **13.** Verfahren nach Anspruch 10, wobei der Schritt des Enterregens der Pumpenmittel den Schritt des Vergleichens der detektieren Spannung mit einer Bezugsspannung aufweist.
- ²⁰ 14. Verfahren nach Anspruch 10, wobei das Verfahren ferner den Schritt des Enterregens des Gebläses aufweist, wenn der detektierte Spannungsabfall größer ist als ein zweiter vorbestimmter Wert, der geringer ist als der erste vorbestimmte Wert.

²⁵ Revendications

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- 1. Système de coupure de compresseur et de cyclage de ventilateur de condenseur permettant de commander un réfrigérant entre un échangeur thermique exothermique et un échangeur thermique endothermique, comprenant :
 - un moyen de pompe actionnable suite à une excitation pour comprimer ledit réfrigérant et pour décharger celui-ci sous pression ;

un moyen de conduit (20) connectant de façon opérationnelle la décharge dudit moyen de pompe au travers dudit échangeur thermique exothermique et dudit échangeur thermique endothermique et un retour sur l'entrée dudit moyen de pompe audit échangeur thermique endothermique ;

un moyen de détente (22) disposé dans ledit moyen de conduit et fonctionnant pour commander l'écoulement d'un réfrigérant entre ledit échangeur thermique exothermique et ledit échangeur thermique endothermique ;

un moyen de détection de température disposé dans ledit conduit entre ledit échangeur thermique exothermique et ledit moyen de détente pour détecter la température dudit réfrigérant sur le côté haute pression dudit moyen de détente ;

un moyen de ventilateur pouvant fonctionner suite à une excitation électrique afin de diriger un écoulement d'air de refroidissement sur ledit échangeur thermique exothermique ; et

un moyen de contrôleur (50) pouvant fonctionner en réponse à ladite détection de température pour exciter ledit moyen de ventilateur lorsque ladite température détectée est supérieure à un premier niveau prédéterminé et pouvant fonctionner pour désexciter ledit moyen de pompe lorsque ladite température détectée est supérieure à un second niveau prédéterminé,

caractérisé en ce que ledit moyen de détection de température comprend une thermistance de détection de température de saturation ;

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un moyen de résistance est monté électriquement en série avec ladite thermistance ;

un moyen de circuit peut fonctionner pour réaliser une circulation de courant au travers dudit moyen de résistance et de ladite thermistance et incluant un moyen qui peut fonctionner pour détecter la chute de tension aux bornes dudit moyen de résistance suite à ladite circulation de courant ; et ledit moyen de contrôleur (50) peut fonctionner en réponse à ladite détection de tension.

- ⁵⁵ 2. Système selon la revendication 1, dans lequel ledit moyen de détente (22) comprend une vanne de détente activée électriquement.
 - 3. Système selon la revendication 1, dans lequel ledit moyen de détente comprend un tube capillaire.

- 4. Système selon la revendication 1, dans lequel ledit moyen de détente comprend une vanne de commande activée par un diaphragme sensible à la température et à la pression.
- 5. Système selon la revendication 1, dans lequel ledit premier niveau de tension prédéterminé est établi à environ 6,6 volts.
- 6. Système selon la revendication 1, dans lequel ledit second niveau prédéterminé est établi à environ 7,3 volts.
- Système selon la revendication 1, dans lequel ledit moyen de contrôleur (50) fonctionne en outre pour désexciter ledit moyen de ventilateur lorsque ladite chute de tension aux bornes dudit moyen de résistance est inférieure audit premier niveau d'une différence prédéterminée.
 - 8. Système selon la revendication 1, dans lequel ledit moyen de contrôleur (50) fonctionne en outre pour ré-exciter ledit moyen de pompe un intervalle temporel prédéterminé après désexcitation.
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9. Système selon la revendication 1, comprenant en outre :

(a) une seconde thermistance disposée dans ledit conduit généralement à la même position que ladite thermistance de détection de température de saturation ;

(b) un moyen de circuit pour produire un signal de tension indicatif de la variation de la résistance de la dite seconde thermistance en réponse à des variations de la température de saturation réelle dudit réfrigérant ; et

(c) un moyen qui peut fonctionner pour comparer ledit second signal de tension de thermistance à ladite tension détectée et qui peut désexciter ledit moyen de pompe lorsque ladite comparaison excède une troisième valeur prédéterminée ou est inférieure à une quatrième valeur prédéterminée.

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10. Procédé de commande de système de réfrigération du type comportant une pompe qui peut fonctionner suite à une excitation pour faire circuler un réfrigérant entre un échangeur thermique exothermique et un échangeur thermique endothermique et un ventilateur pour diriger un écoulement d'air sur ledit échangeur thermique exothermique, comprenant les étapes de :

- (a) fourniture d'un moyen de détente (22, 222) pour commander un écoulement du réfrigérant depuis ledit échangeur thermique exothermique jusqu'audit échangeur thermique endothermique ; et
 (b) fourniture d'un détecteur (Ts) pour détecter la température du réfrigérant qui pénètre dans ledit moyen de détente et pour connecter une résistance électriquement en série avec ledit détecteur ;
 (c) circulation d'un courant électrique au travers de ladite résistance et dudit détecteur ;
- (d) détection de la chute de tension aux bornes de ladite résistance lorsque ledit courant circule ;
 (e) excitation d'un ventilateur (32) et direction d'un écoulement d'air de refroidissement sur ledit échangeur thermique exothermique lorsque ladite chute de tension détectée est supérieure à une première valeur prédéterminée et

(f) désexcitation dudit moyen de pompe lorsque ladite chute de tension détectée est supérieure à une seconde valeur prédéterminée.

- **11.** Procédé selon la revendication 10, dans lequel ladite étape de fourniture d'un détecteur inclut l'étape de fourniture d'une thermistance (Ts) dans l'écoulement du réfrigérant jusqu'audit moyen de détente.
- **12.** Procédé selon la revendication 10, dans lequel ladite étape d'excitation dudit ventilateur inclut l'étape de comparaison de ladite tension détectée à une tension de référence.
 - **13.** Procédé selon la revendication 10, dans lequel ladite étape de désexcitation dudit moyen de pompe inclut l'étape de comparaison de ladite tension détectée à une valeur de référence.
- ⁵⁰ 14. Procédé selon la revendication 10, comprenant en outre l'étape de désexcitation dudit ventilateur lorsque ladite chute de tension détectée est supérieure à une seconde valeur prédéterminée inférieure à ladite première valeur prédéterminée.











